

Bioengineering 280A  
Principles of Biomedical Imaging

Fall Quarter 2012  
MRI Lecture 3

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### Precession

$$\begin{bmatrix} dM_x/dt \\ dM_y/dt \\ dM_z/dt \end{bmatrix} = \gamma \begin{bmatrix} 0 & B_0 & 0 \\ -B_0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} M_x \\ M_y \\ M_z \end{bmatrix}$$

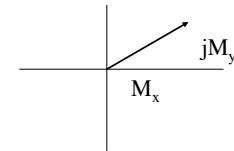
Useful to define  $M \equiv M_x + jM_y$

$$\begin{aligned} dM/dt &= d/dt(M_x + jM_y) \\ &= -j\gamma B_0 M \end{aligned}$$

Solution is a time-varying phasor

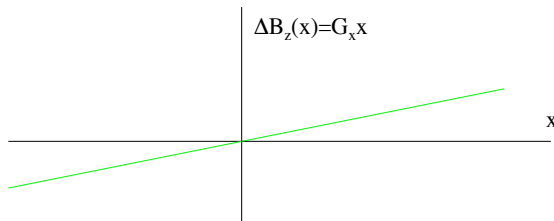
$$M(t) = M(0)e^{-j\gamma B_0 t} = M(0)e^{-j\omega_0 t}$$

Question: which way does this rotate with time?



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### Interpretation



Spins Precess at  $\gamma B_0 - \gamma G_x x$  (slower)

$$\begin{aligned} M(t) &= M(0)e^{-j\gamma(B_0 - G_x x)t} \\ &= M(0)e^{-j(\omega_0 - \Delta\omega)t} \end{aligned}$$

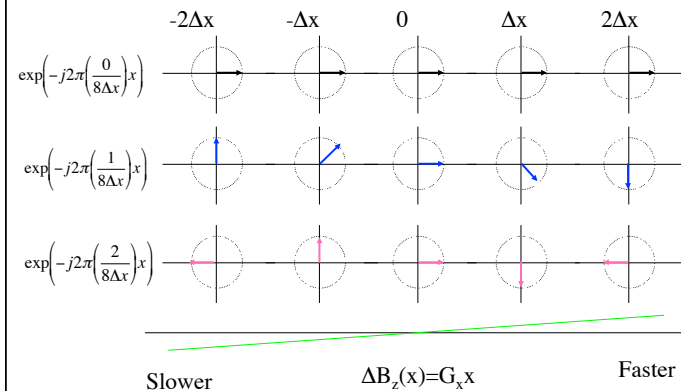
Spins Precess at  $\gamma B_0$   
 $M(t) = M(0)e^{-j\gamma B_0 t}$   
 $= M(0)e^{-j\omega_0 t}$

Spins Precess at  $\gamma B_0 + \gamma G_x x$  (faster)

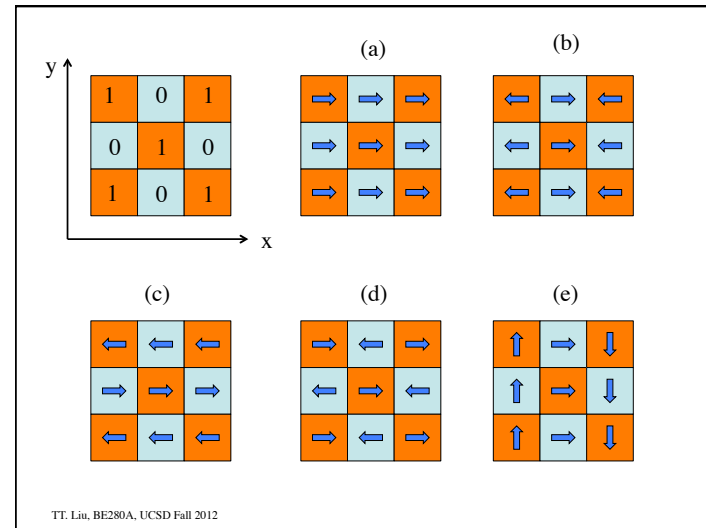
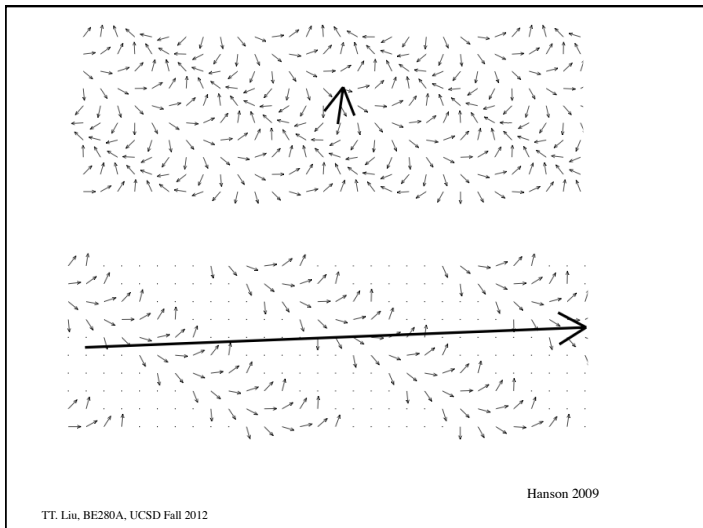
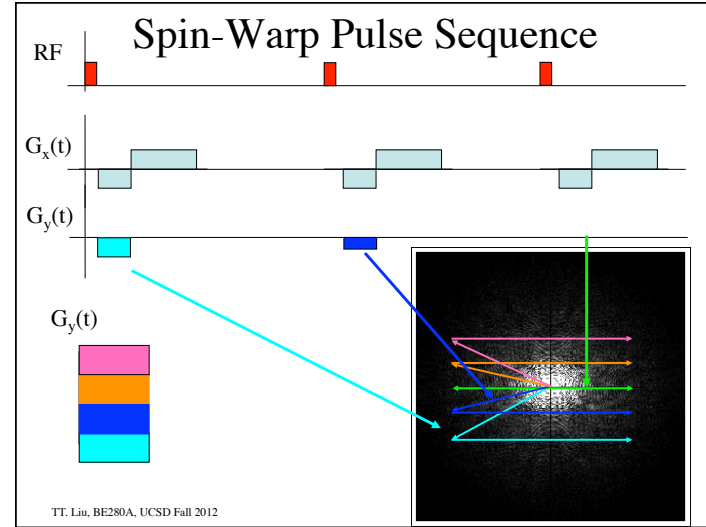
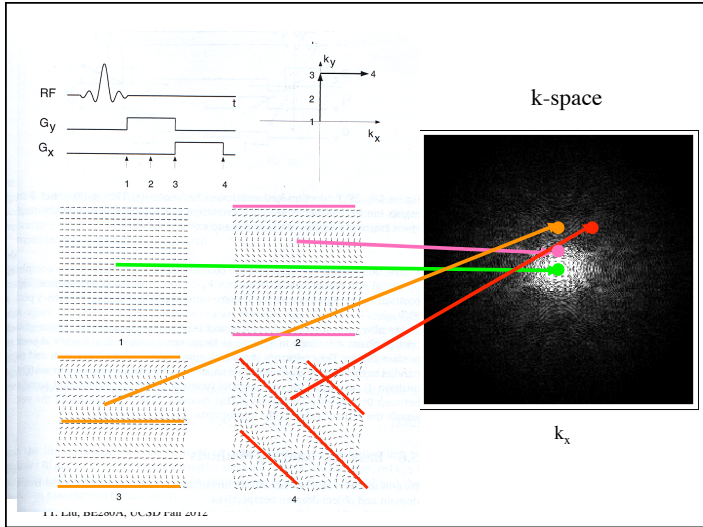
$$\begin{aligned} M(t) &= M(0)e^{-j\gamma(B_0 + G_x x)t} \\ &= M(0)e^{-j(\omega_0 + \Delta\omega)t} \end{aligned}$$

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### Interpretation



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## Gradient Fields

Define

$$\vec{G} \equiv G_x \hat{i} + G_y \hat{j} + G_z \hat{k} \quad \vec{r} \equiv x \hat{i} + y \hat{j} + z \hat{k}$$

So that

$$G_x x + G_y y + G_z z = \vec{G} \cdot \vec{r}$$

Also, let the gradient fields be a function of time. Then the z-directed magnetic field at each point in the volume is given by :

$$B_z(\vec{r}, t) = B_0 + \vec{G}(t) \cdot \vec{r}$$

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## Static Gradient Fields

In a uniform magnetic field, the transverse magnetization is given by:

$$M(t) = M(0) e^{-j\omega_0 t} e^{-t/T_2}$$

In the presence of non time-varying gradients we have

$$\begin{aligned} M(\vec{r}) &= M(\vec{r}, 0) e^{-j\gamma B_z(\vec{r})t} e^{-t/T_2(\vec{r})} \\ &= M(\vec{r}, 0) e^{-j\gamma(B_0 + \vec{G} \cdot \vec{r})t} e^{-t/T_2(\vec{r})} \\ &= M(\vec{r}, 0) e^{-j\omega_0 t} e^{-j\gamma \vec{G} \cdot \vec{r} t} e^{-t/T_2(\vec{r})} \end{aligned}$$

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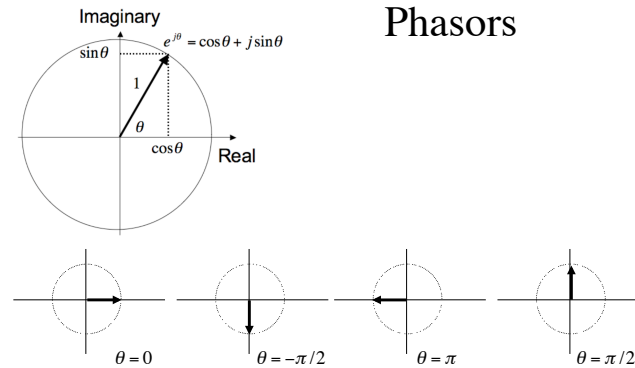
## Time-Varying Gradient Fields

In the presence of time-varying gradients the frequency as a function of space and time is:

$$\begin{aligned} \omega(\vec{r}, t) &= \gamma B_z(\vec{r}, t) \\ &= \gamma B_0 + \gamma \vec{G}(t) \cdot \vec{r} \\ &= \omega_0 + \Delta\omega(\vec{r}, t) \end{aligned}$$

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## Phasors



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## Phase

Phase = angle of the magnetization phasor  
 Frequency = rate of change of angle (e.g. radians/sec)  
 Phase = time integral of frequency

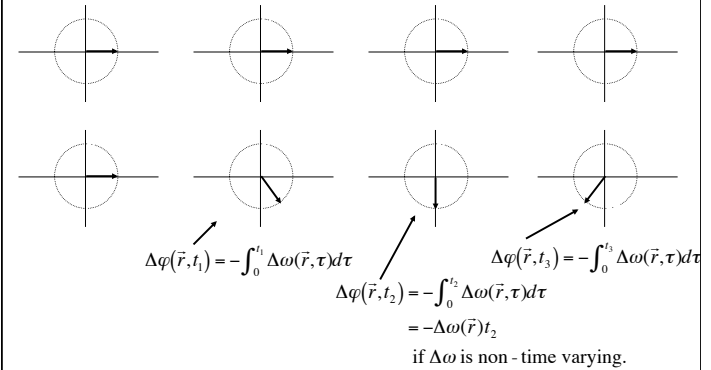
$$\begin{aligned}\varphi(\vec{r}, t) &= -\int_0^t \omega(\vec{r}, \tau) d\tau \\ &= -\omega_0 t + \Delta\varphi(\vec{r}, t)\end{aligned}$$

Where the incremental phase due to the gradients is

$$\begin{aligned}\Delta\varphi(\vec{r}, t) &= -\int_0^t \Delta\omega(\vec{r}, \tau) d\tau \\ &= -\int_0^t \gamma \vec{G}(\tau) \cdot \vec{r} d\tau\end{aligned}$$

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## Phase with constant gradient



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## Time-Varying Gradient Fields

The transverse magnetization is then given by

$$\begin{aligned}M(\vec{r}, t) &= M(\vec{r}, 0) e^{-t/T_2(\vec{r})} e^{j\varphi(\vec{r}, t)} \\ &= M(\vec{r}, 0) e^{-t/T_2(\vec{r})} e^{-j\omega_0 t} \exp\left(-j \int_0^t \Delta\omega(\vec{r}, \tau) d\tau\right) \\ &= M(\vec{r}, 0) e^{-t/T_2(\vec{r})} e^{-j\omega_0 t} \exp\left(-j\gamma \int_0^t \vec{G}(\tau) \cdot \vec{r} d\tau\right)\end{aligned}$$

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## Signal Equation

Signal from a volume

$$\begin{aligned}s_r(t) &= \int_V M(\vec{r}, t) dV \\ &= \int_x \int_y \int_z M(x, y, z, 0) e^{-t/T_2(\vec{r})} e^{-j\omega_0 t} \exp\left(-j\gamma \int_0^t \vec{G}(\tau) \cdot \vec{r} d\tau\right) dx dy dz\end{aligned}$$

For now, consider signal from a slice along  $z$  and drop the  $T_2$  term. Define  $m(x, y) \equiv \int_{z_0 - \Delta z/2}^{z_0 + \Delta z/2} M(\vec{r}, t) dz$

To obtain

$$s_r(t) = \int_x \int_y m(x, y) e^{-j\omega_0 t} \exp\left(-j\gamma \int_0^t \vec{G}(\tau) \cdot \vec{r} d\tau\right) dx dy$$

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## Signal Equation

Demodulate the signal to obtain

$$\begin{aligned}
 s(t) &= e^{j\omega_0 t} s_r(t) \\
 &= \int_x \int_y m(x, y) \exp\left(-j\gamma \int_0^t \vec{G}(\tau) \cdot \vec{r} d\tau\right) dx dy \\
 &= \int_x \int_y m(x, y) \exp\left(-j\gamma \int_0^t [G_x(\tau)x + G_y(\tau)y] d\tau\right) dx dy \\
 &= \int_x \int_y m(x, y) \exp\left(-j2\pi(k_x(t)x + k_y(t)y)\right) dx dy
 \end{aligned}$$

Where

$$\begin{aligned}
 k_x(t) &= \frac{\gamma}{2\pi} \int_0^t G_x(\tau) d\tau \\
 k_y(t) &= \frac{\gamma}{2\pi} \int_0^t G_y(\tau) d\tau
 \end{aligned}$$

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## MR signal is Fourier Transform

$$\begin{aligned}
 s(t) &= \int_x \int_y m(x, y) \exp\left(-j2\pi(k_x(t)x + k_y(t)y)\right) dx dy \\
 &= M(k_x(t), k_y(t)) \\
 &= F[m(x, y)]_{k_x(t), k_y(t)}
 \end{aligned}$$

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## Recap

- Frequency = rate of change of phase.
- Higher magnetic field -> higher Larmor frequency -> phase changes more rapidly with time.
- With a constant gradient  $G_x$ , spins at different x locations precess at different frequencies -> spins at greater x-values change phase more rapidly.
- With a constant gradient, distribution of phases across x locations changes with time. (phase modulation)
- More rapid change of phase with x -> higher spatial frequency  $k_x$

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## K-space

At each point in time, the received signal is the Fourier transform of the object

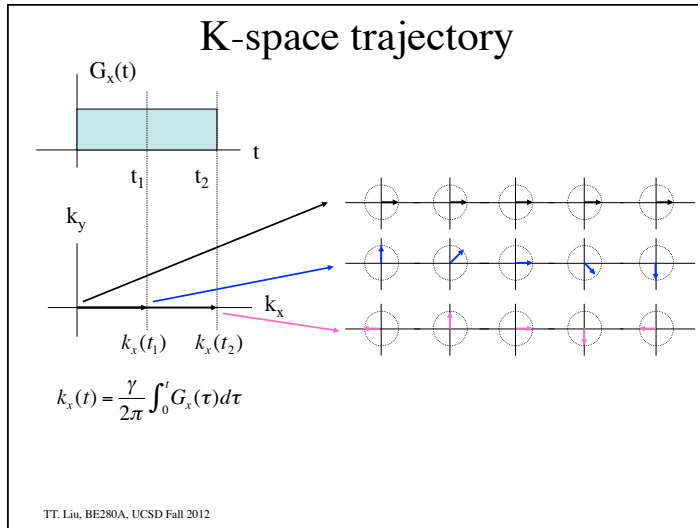
$$s(t) = M(k_x(t), k_y(t)) = F[m(x, y)]_{k_x(t), k_y(t)}$$

evaluated at the spatial frequencies:

$$\begin{aligned}
 k_x(t) &= \frac{\gamma}{2\pi} \int_0^t G_x(\tau) d\tau \\
 k_y(t) &= \frac{\gamma}{2\pi} \int_0^t G_y(\tau) d\tau
 \end{aligned}$$

Thus, the gradients control our position in k-space. The design of an MRI pulse sequence requires us to efficiently cover enough of k-space to form our image.

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### Units

Spatial frequencies ( $k_x, k_y$ ) have units of 1/distance.  
Most commonly, 1/cm

Gradient strengths have units of (magnetic field)/  
distance. Most commonly G/cm or mT/m

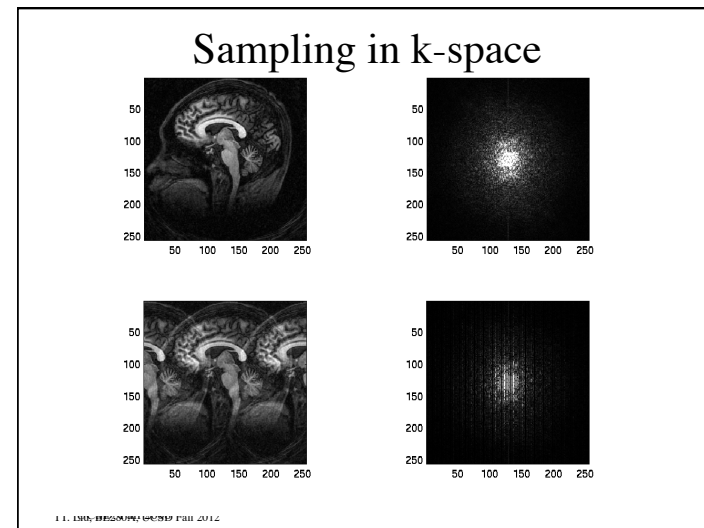
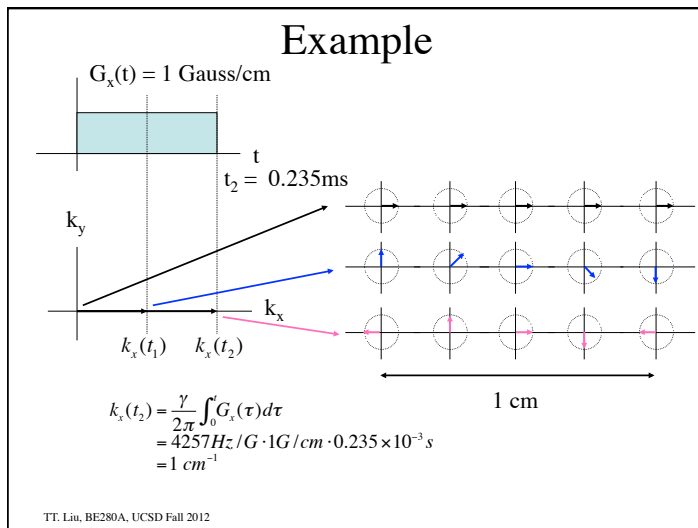
$\gamma/(2\pi)$  has units of Hz/G or Hz/Tesla.

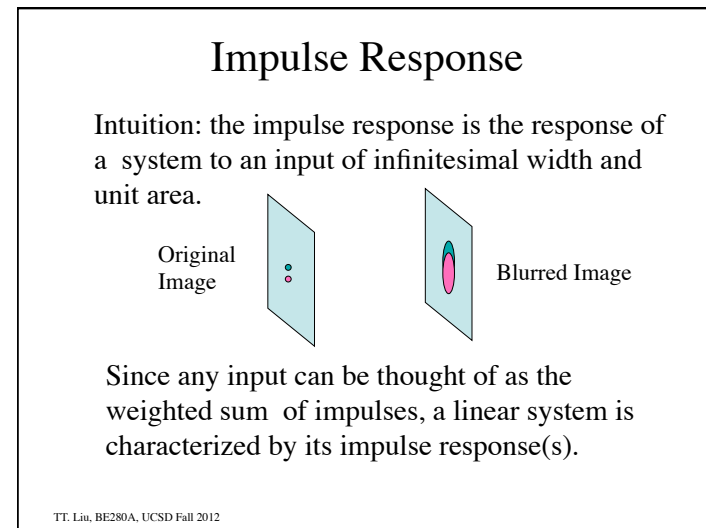
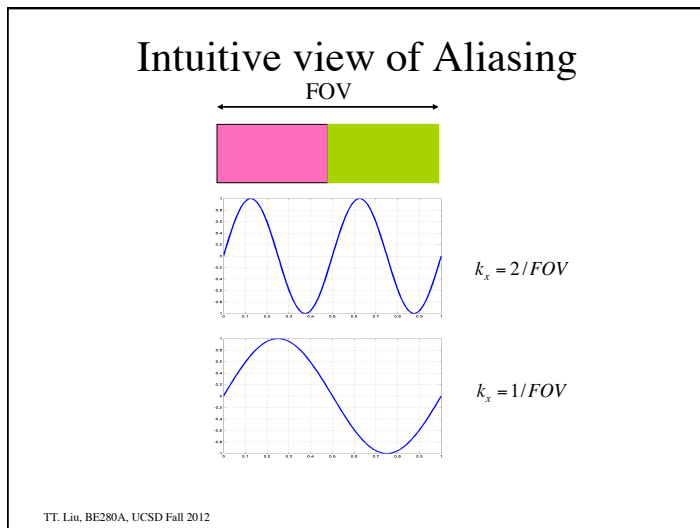
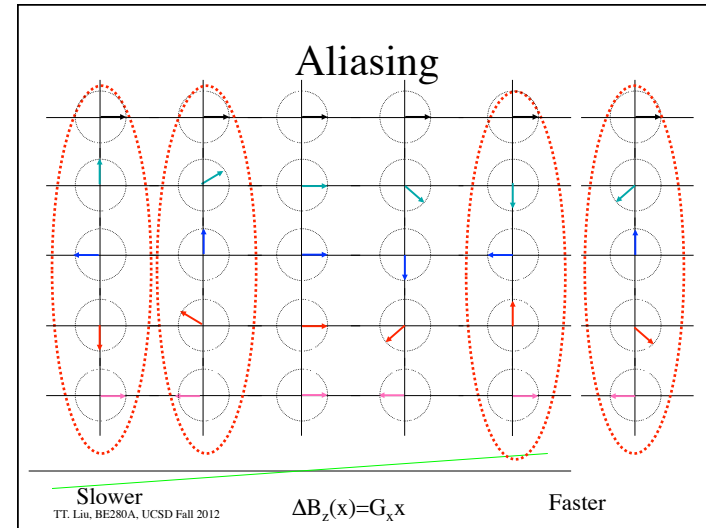
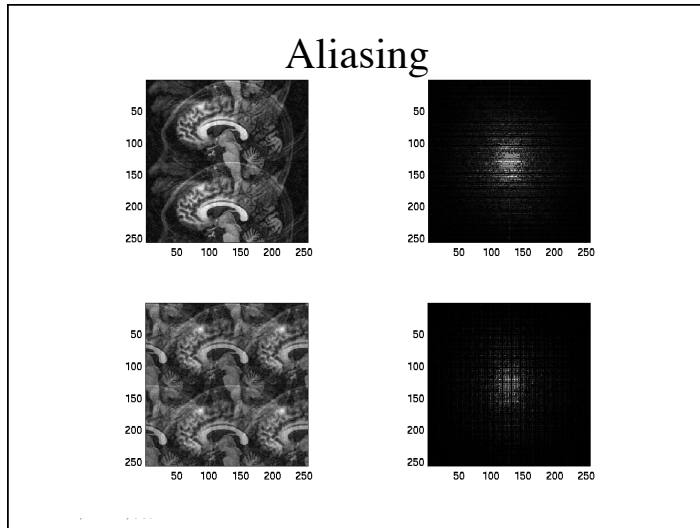
$$k_x(t) = \frac{\gamma}{2\pi} \int_0^t G_x(\tau) d\tau$$

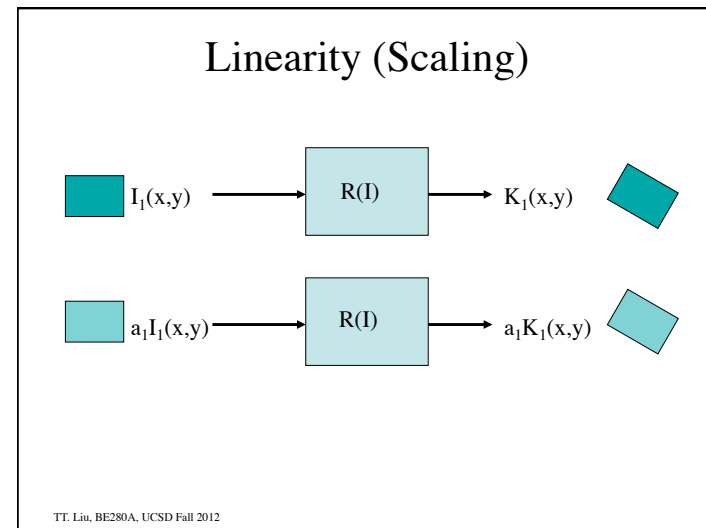
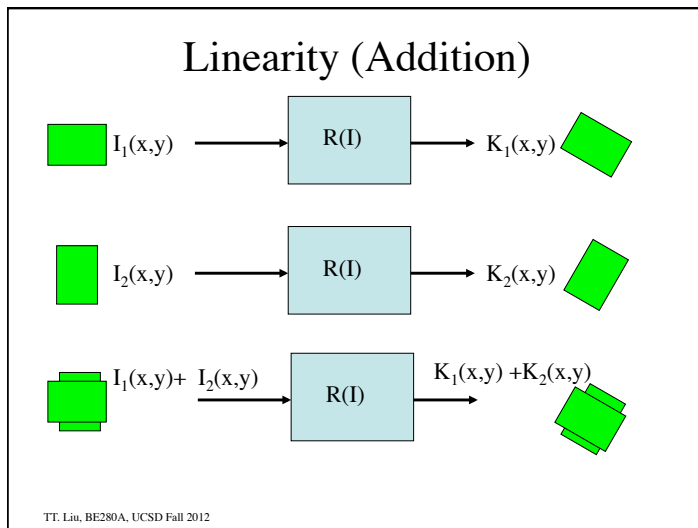
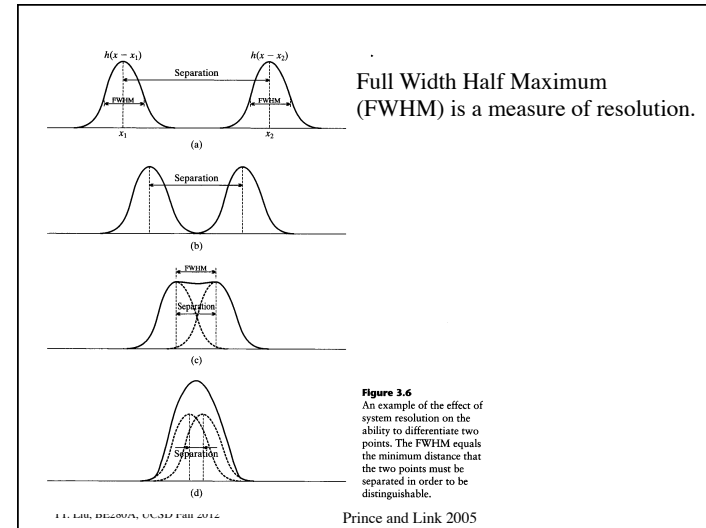
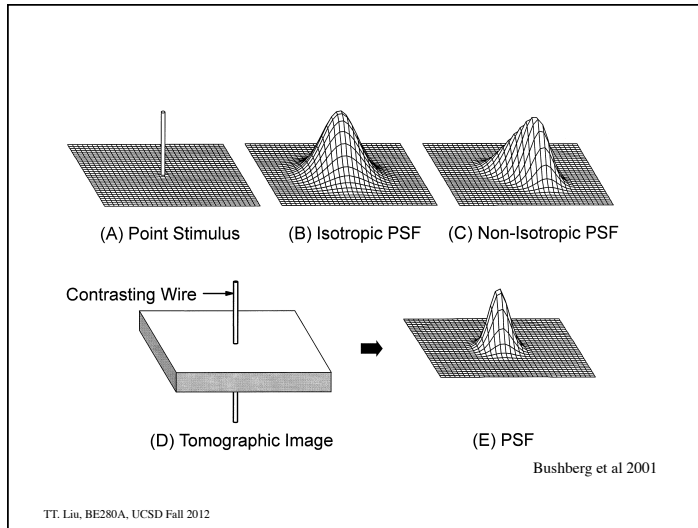
$$= [\text{Hz/Gauss}][\text{Gauss/cm}][\text{sec}]$$

$$= [1/\text{cm}]$$

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## Linearity

A system  $R$  is linear if for two inputs  $I_1(x,y)$  and  $I_2(x,y)$  with outputs

$$R(I_1(x,y))=K_1(x,y) \text{ and } R(I_2(x,y))=K_2(x,y)$$

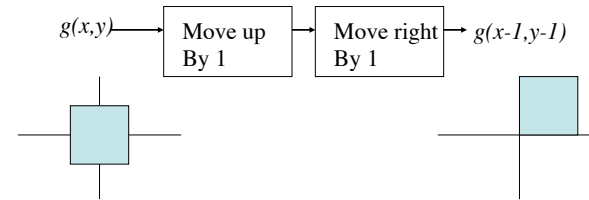
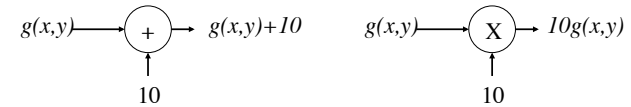
the response to the weighted sum of inputs is the weighted sum of outputs:

$$R(a_1I_1(x,y) + a_2I_2(x,y))=a_1K_1(x,y) + a_2K_2(x,y)$$

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## Example

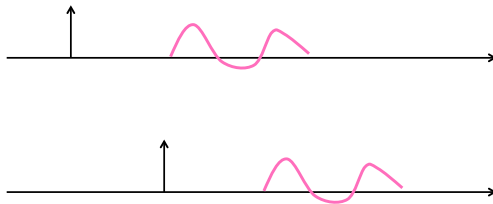
Are these linear systems?



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## Time Invariance

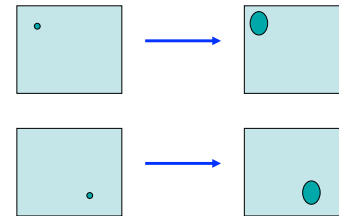
If a system is time invariant, the impulse response depends only on the difference between the time of the output and the position of the impulse and is given by  $h(t_2;t) = h(t_2 - t)$



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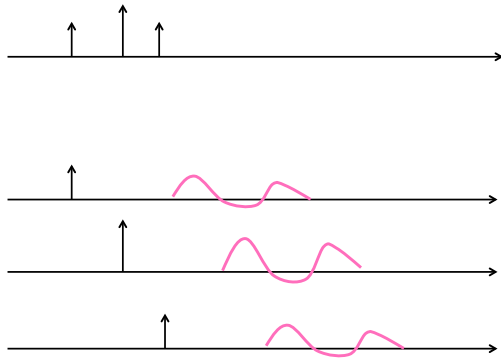
## Space Invariance

If a system is space invariant, the impulse response depends only on the difference between the output coordinates and the position of the impulse and is given by  $h(x_2,y_2;\xi,\eta) = h(x_2 - \xi, y_2 - \eta)$



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## LTI Response



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## 1D Convolution

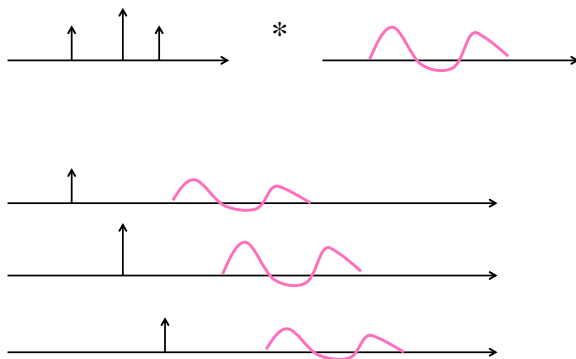
$$\begin{aligned} I(x) &= \int_{-\infty}^{\infty} g(\xi)h(x;\xi)d\xi \\ &= \int_{-\infty}^{\infty} g(\xi)h(x-\xi)d\xi \\ &= g(x) * h(x) \end{aligned}$$

Useful fact:

$$\begin{aligned} g(x) * \delta(x-\Delta) &= \int_{-\infty}^{\infty} g(\xi)\delta(x-\Delta-\xi)d\xi \\ &= g(x-\Delta) \end{aligned}$$

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## LTI Response



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## 2D Convolution

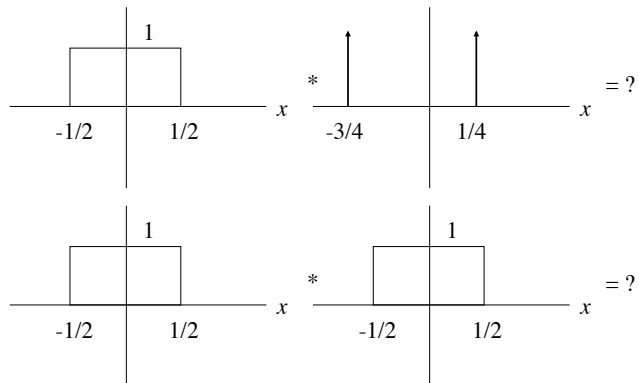
For a space invariant linear system, the superposition integral becomes a convolution integral.

$$\begin{aligned} I(x_2, y_2) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} g(\xi, \eta)h(x_2, y_2; \xi, \eta)d\xi d\eta \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} g(\xi, \eta)h(x_2 - \xi, y_2 - \eta)d\xi d\eta \\ &= g(x_2, y_2) ** h(x_2, y_2) \end{aligned}$$

where \*\* denotes 2D convolution. This will sometimes be abbreviated as \*, e.g.  $I(x_2, y_2) = g(x_2, y_2) * h(x_2, y_2)$ .

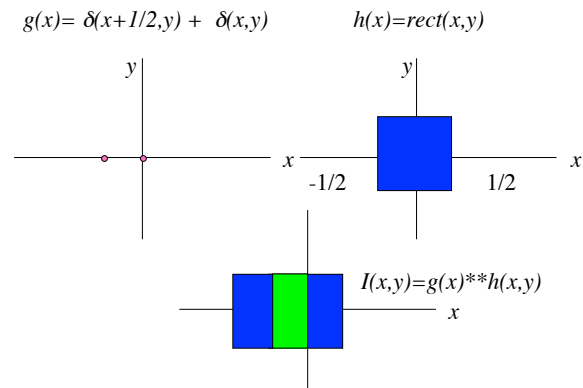
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## 1D Convolution Examples



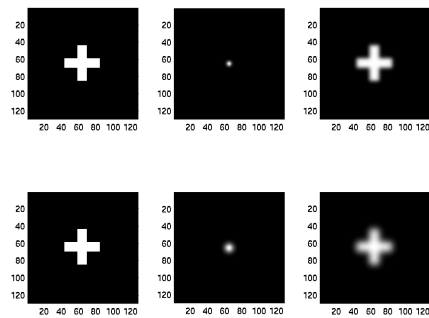
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## 2D Convolution Example



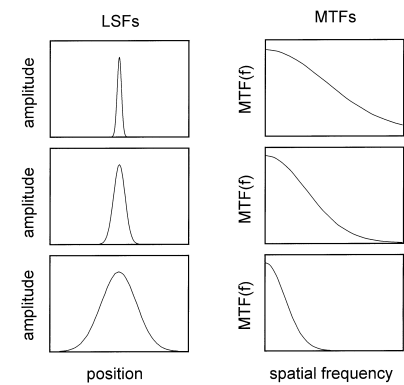
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## 2D Convolution Example



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## MTF = Fourier Transform of Impulse Response



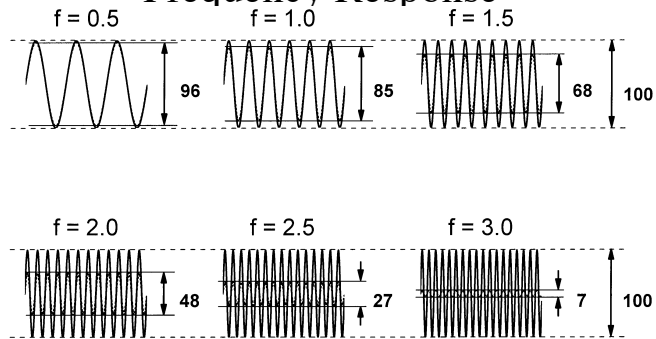
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Bushberg et al 2001

## Modulation Transfer Function (MTF)

or

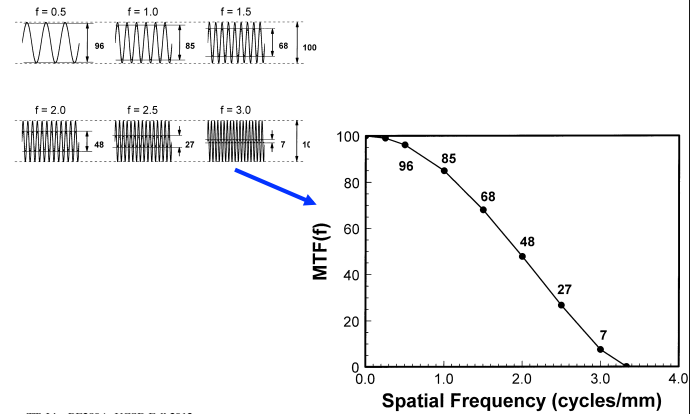
## Frequency Response



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Bushberg et al 2001

## Modulation Transfer Function



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Bushberg et al 2001

Figure 1:

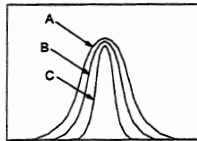
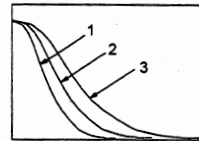


Figure 2:



8. Referring to Figure 1 (above) which demonstrates 3 different line spread functions (LSF), which LSF will yield the best spatial resolution?
10. Referring to Figure 1 which shows LSFs, and Figure 2 which shows the corresponding modulation transfer functions (MTFs), which MTF corresponds to LSF C?
- A. MTF number 1  
B. MTF number 2  
C. MTF number 3

- D74. The intrinsic resolution of a gamma camera is 5 mm. The collimator resolution is 10 mm. The overall system resolution is \_\_\_\_\_ mm.
- A. 15  
B. 11.2  
C. 7.5  
D. 5.0  
E. 0.5

## Eigenfunctions

The fundamental nature of the convolution theorem may be better understood by observing that the complex exponentials are eigenfunctions of the convolution operator.

$$e^{j2\pi k_x x} \longrightarrow \boxed{g(x)} \longrightarrow z(x)$$

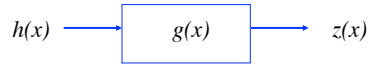
$$\begin{aligned} z(x) &= g(x) * e^{j2\pi k_x x} \\ &= \int_{-\infty}^{\infty} g(u) e^{j2\pi k_x (x-u)} du \\ &= G(k_x) e^{j2\pi k_x x} \end{aligned}$$

The response of a linear shift invariant system to a complex exponential is simply the exponential multiplied by the FT of the system's impulse response.

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## Convolution/Multiplication

Now consider an arbitrary input  $h(x)$ .



Recall that we can express  $h(x)$  as the integral of weighted complex exponentials.

$$h(x) = \int_{-\infty}^{\infty} H(k_x) e^{j2\pi k_x x} dk_x$$

Each of these exponentials is weighted by  $G(k_x)$  so that the response may be written as

$$z(x) = \int_{-\infty}^{\infty} G(k_x) H(k_x) e^{j2\pi k_x x} dk_x$$

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## Convolution/Modulation Theorem

$$\begin{aligned} F\{g(x) * h(x)\} &= \int_{-\infty}^{\infty} \left[ \int_{-\infty}^{\infty} g(u) * h(x-u) du \right] e^{-j2\pi k_x x} dx \\ &= \int_{-\infty}^{\infty} g(u) \int_{-\infty}^{\infty} h(x-u) e^{-j2\pi k_x x} dx du \\ &= \int_{-\infty}^{\infty} g(u) H(k_x) e^{-j2\pi k_x u} du \\ &= G(k_x) H(k_x) \end{aligned}$$

Convolution in the spatial domain transforms into multiplication in the frequency domain. Dual is modulation

$$F\{g(x)h(x)\} = G(k_x) * H(k_x)$$

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## 2D Convolution/Multiplication

*Convolution*

$$F[g(x,y) * h(x,y)] = G(k_x, k_y) H(k_x, k_y)$$

*Multiplication*

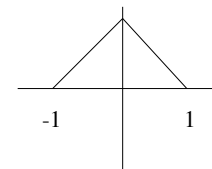
$$F[g(x,y)h(x,y)] = G(k_x, k_y) * H(k_x, k_y)$$

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## Application of Convolution Thm.

$$\Lambda(x) = \begin{cases} 1 - |x| & |x| < 1 \\ 0 & \text{otherwise} \end{cases}$$

$$F(\Lambda(x)) = \int_{-1}^1 (1 - |x|) e^{-j2\pi k_x x} dx = ??$$

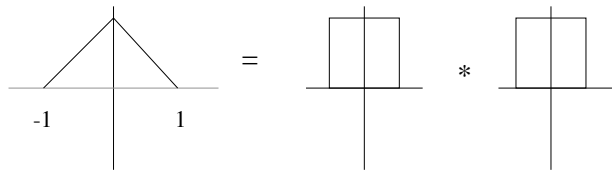


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## Application of Convolution Thm.

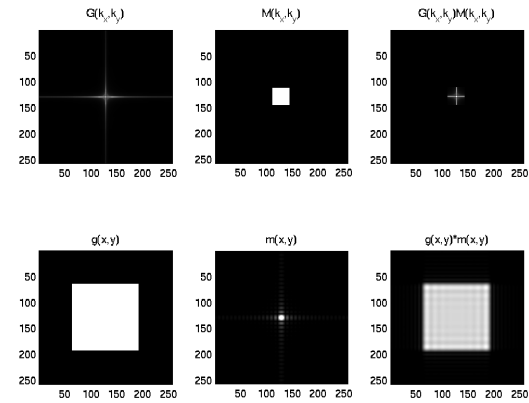
$$\Lambda(x) = \Pi(x) * \Pi(x)$$

$$F(\Lambda(x)) = \text{sinc}^2(k_x)$$

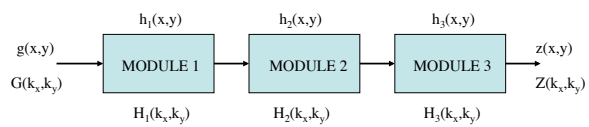


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## Convolution Example



## Response of an Imaging System

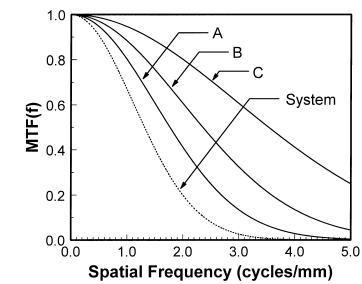


$$z(x, y) = g(x, y) ** h_1(x, y) ** h_2(x, y) ** h_3(x, y)$$

$$Z(k_x, k_y) = G(k_x, k_y) H_1(k_x, k_y) H_2(k_x, k_y) H_3(k_x, k_y)$$

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## System MTF = Product of MTFs of Components



Bushberg et al 2001

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Useful Approximation

$$FWHM_{System} = \sqrt{FWHM_1^2 + FWHM_2^2 + \dots + FWHM_N^2}$$

Example

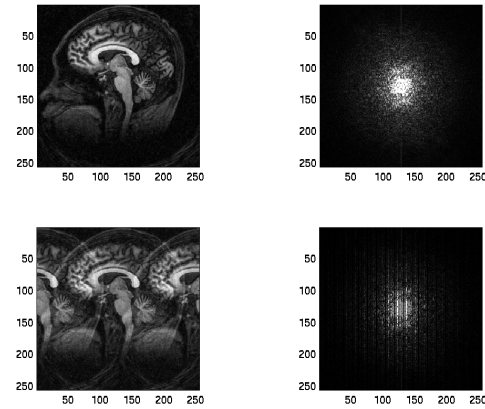
$$FWHM_1 = 1mm$$

$$FWHM_2 = 2mm$$

$$FWHM_{system} = \sqrt{5} = 2.24mm$$

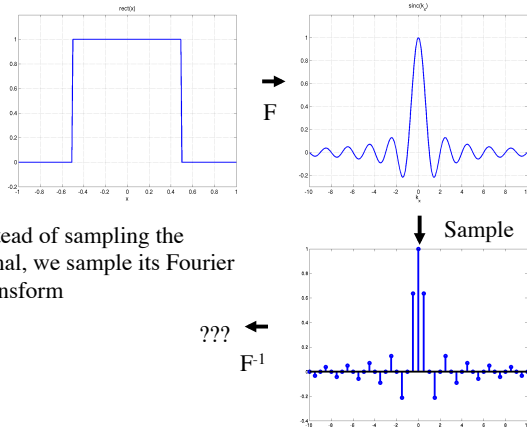
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## Sampling in k-space



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## Fourier Sampling

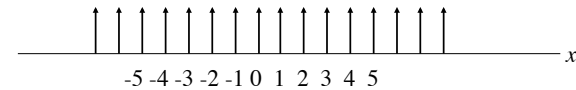


Instead of sampling the signal, we sample its Fourier Transform

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## Comb Function

$$comb(x) = \sum_{n=-\infty}^{\infty} \delta(x - n)$$

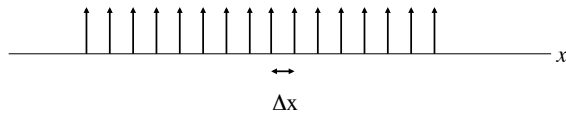


Other names: Impulse train, bed of nails, shah function.

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### Scaled Comb Function

$$\begin{aligned} \text{comb}\left(\frac{x}{\Delta x}\right) &= \sum_{n=-\infty}^{\infty} \delta\left(\frac{x}{\Delta x} - n\right) \\ &= \sum_{n=-\infty}^{\infty} \delta\left(\frac{x - n\Delta x}{\Delta x}\right) \\ &= \Delta x \sum_{n=-\infty}^{\infty} \delta(x - n\Delta x) \end{aligned}$$



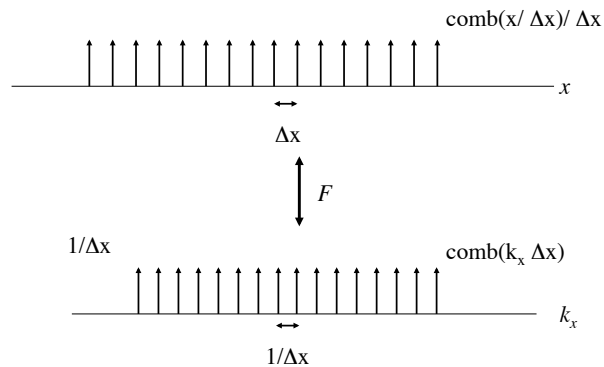
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### Fourier Transform of comb(x)

$$\begin{aligned} F[\text{comb}(x)] &= \text{comb}(k_x) \\ &= \sum_{n=-\infty}^{\infty} \delta(k_x - n) \\ F\left[\frac{1}{\Delta x} \text{comb}\left(\frac{x}{\Delta x}\right)\right] &= \frac{1}{\Delta x} \Delta x \text{comb}(k_x \Delta x) \\ &= \sum_{n=-\infty}^{\infty} \delta(k_x \Delta x - n) \\ &= \frac{1}{\Delta x} \sum_{n=-\infty}^{\infty} \delta\left(k_x - \frac{n}{\Delta x}\right) \end{aligned}$$

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### Fourier Transform of comb(x/ Δx)



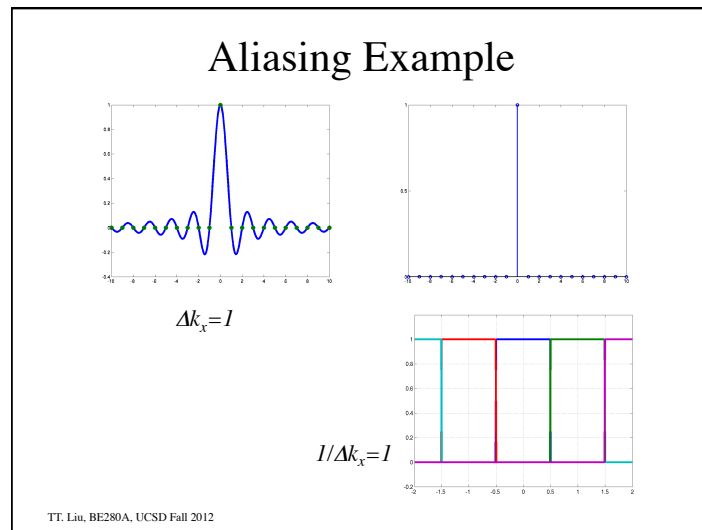
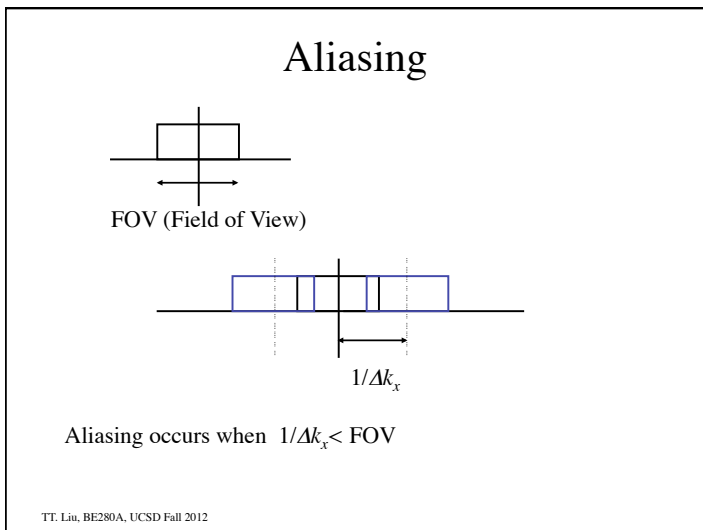
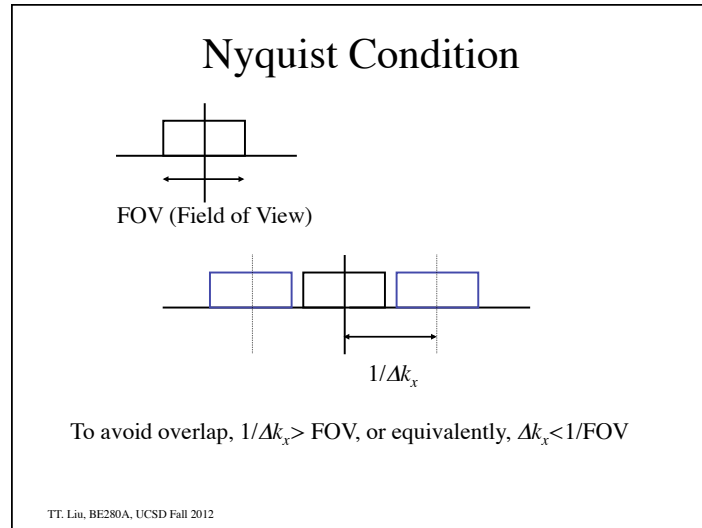
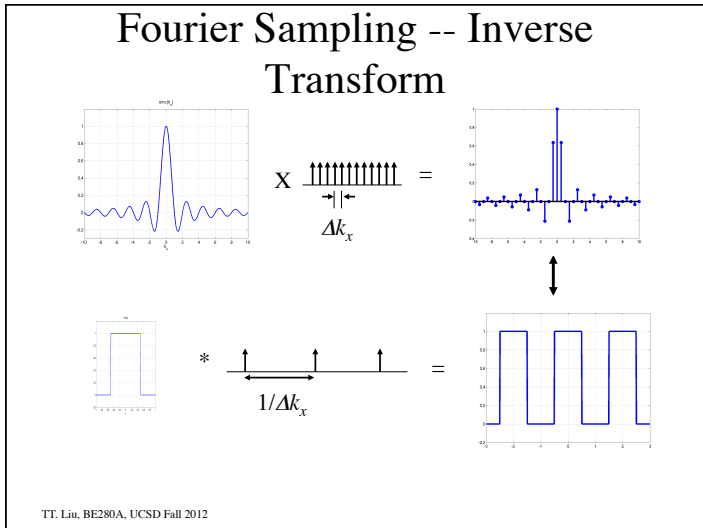
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### Fourier Sampling

$$\begin{aligned} &(1/\Delta k_x) \text{comb}(k_x/\Delta k_x) \\ G_S(k_x) &= G(k_x) \frac{1}{\Delta k_x} \text{comb}\left(\frac{k_x}{\Delta k_x}\right) \\ &= G(k_x) \sum_{n=-\infty}^{\infty} \delta(k_x - n\Delta k_x) \\ &= \sum_{n=-\infty}^{\infty} G(n\Delta k_x) \delta(k_x - n\Delta k_x) \end{aligned}$$

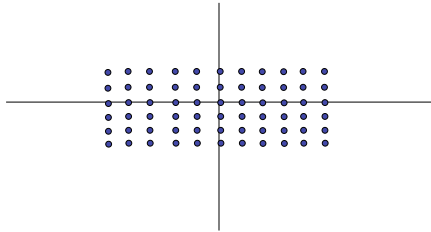
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## 2D Comb Function

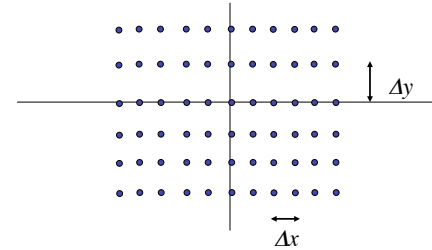
$$\begin{aligned} \text{comb}(x, y) &= \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} \delta(x - m, y - n) \\ &= \sum_{m=-\infty}^{\infty} \delta(x - m) \sum_{n=-\infty}^{\infty} \delta(y - n) \\ &= \text{comb}(x) \text{comb}(y) \end{aligned}$$



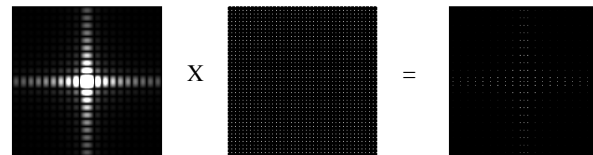
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## Scaled 2D Comb Function

$$\begin{aligned} \text{comb}(x/\Delta x, y/\Delta y) &= \text{comb}(x/\Delta x) \text{comb}(y/\Delta y) \\ &= \Delta x \Delta y \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} \delta(x - m\Delta x) \delta(y - n\Delta y) \end{aligned}$$



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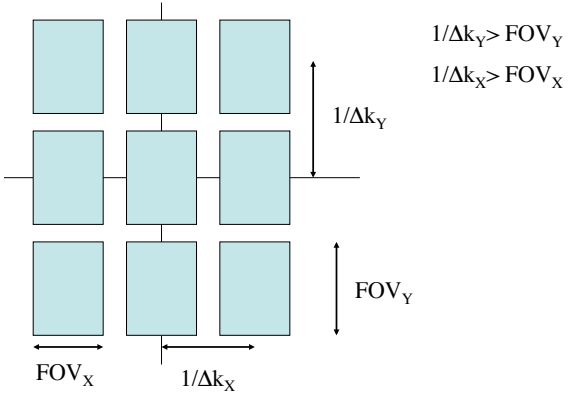
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## 2D k-space sampling

$$\begin{aligned} G_S(k_x, k_y) &= G(k_x, k_y) \frac{1}{\Delta k_x \Delta k_y} \text{comb}\left(\frac{k_x}{\Delta k_x}, \frac{k_y}{\Delta k_y}\right) \\ &= G(k_x, k_y) \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} \delta(k_x - m\Delta k_x, k_y - n\Delta k_y) \\ &= \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} G(m\Delta k_x, n\Delta k_y) \delta(k_x - m\Delta k_x, k_y - n\Delta k_y) \end{aligned}$$

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# Nyquist Conditions



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